Physics 696 Topics in Advanced Accelerator Design I Monday, September 10 2012

E&M: Plane Waves, Waveguides, Boundary Absorption

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Happy Birthday to Arthur Compton (1927 Nobel), Joey Votto, Colin Firth, and the LHC! Happy Swap Ideas Day, International Make-Up Day, and Cheap Advice Day!

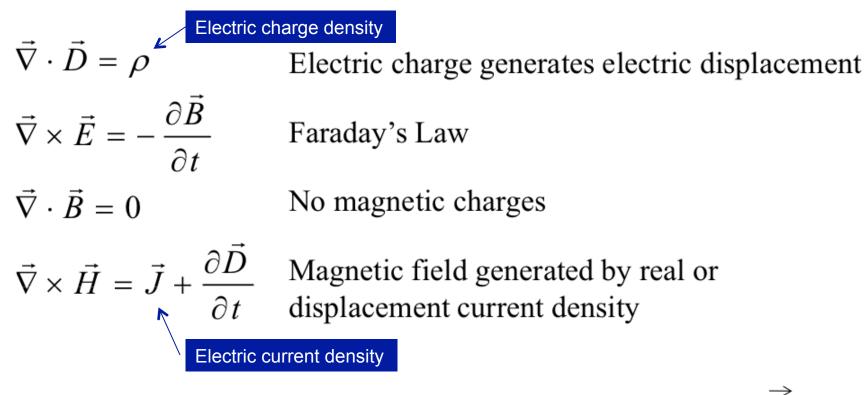


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Maxwell's Equations



Maxwell's Equations are linear in the source terms ρ and \overline{J} . In general will generate linear (partial) differential equations to solve. Superposition valid!

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Constitutive Relations

 \vec{D} : Electric displacement field

$$\vec{D} = \varepsilon \vec{E} = \varepsilon_r \varepsilon_0 \vec{E}$$

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$$\varepsilon_0 = 8.85 \times 10^{-12} \frac{\mathrm{C}}{\mathrm{N} \cdot \mathrm{m}^2}$$

$$arepsilon$$
 : $ec{B}=\muec{H}=\mu_r\mu_0ec{H}$ μ_0 =

$$\mu_0 = 4\pi \times 10^{-7} \, \frac{\mathrm{N} \cdot \mathrm{s}^2}{\mathrm{C}^2}$$

- \vec{B} : Magnetic field
- \vec{H} : Magnetizing field
 - $\varepsilon: \mbox{Permeability}$

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 \vec{E} : Electric field

Permittivity



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 $\varepsilon_0 \mu_0 = \frac{1}{c^2}$

Source-Free Maxwell Equations

• When there are no sources ($\vec{J} = 0, \rho = 0$) then Maxwell's equations become much more symmetric:

$$\vec{\nabla} \cdot \vec{D} = 0$$

$$\vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t} \qquad \vec{D} = \epsilon \vec{E}$$

$$\vec{\nabla} \cdot \vec{B} = 0 \qquad \vec{H} = \frac{1}{\mu} \vec{B}$$

$$\vec{\nabla} \times \vec{H} = \frac{\partial \vec{D}}{\partial t}$$

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Harmonic Dependence of Fields

The linearity of the first order time derivatives suggests that we can assume a harmonic time dependence of our magnetic and electric fields

$$\vec{B}(\vec{x},t) = \vec{B}_0(\vec{x}) e^{-i\omega t}$$
$$\vec{E}(\vec{x},t) = \vec{E}_0(\vec{x}) e^{-i\omega t}$$

This gives the spatial Maxwell Equations:

 $\vec{\nabla} \cdot \vec{B} = 0$ $\vec{\nabla} \times \vec{E} - i\omega \vec{B} = 0$ $\vec{\nabla} \cdot \vec{E} = 0$ $\vec{\nabla} \times \vec{B} + i\omega\mu\epsilon\vec{E} = 0$



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Plane Waves

 Taking the curl of each curl equation and using the identity

$$\vec{\nabla} \times (\vec{\nabla} \times \vec{E}) = \vec{\nabla} (\vec{\nabla} \cdot \vec{E}) - \nabla^2 \vec{E} = -\nabla^2 \vec{E}$$

then gives us two **identical** spatial wave equations with straightforward solutions:

$$[\nabla^2 + \mu \epsilon \omega^2] \vec{E} = 0 \quad \Rightarrow \quad \vec{E}(\vec{x}, t) = \vec{E}_0 e^{i\vec{k}\cdot\vec{x} - i\omega t}$$
$$[\nabla^2 + \mu \epsilon \omega^2] \vec{B} = 0 \quad \Rightarrow \quad \vec{B}(\vec{x}, t) = \vec{B}_0 e^{i\vec{k}\cdot\vec{x} - i\omega t}$$

These are plane waves traveling in direction \vec{k}



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Plane Wave Properties

$$\vec{E}(\vec{x},t) = \vec{E}_0 e^{i\vec{k}\cdot\vec{x}-i\omega t}$$
$$\vec{B}(\vec{x},t) = \vec{B}_0 e^{i\vec{k}\cdot\vec{x}-i\omega t}$$

The source-free divergence equations imply $\vec{k} \cdot \vec{E}_0 = \vec{k} \cdot \vec{B}_0 = 0$ so the fields are both transverse to the direction \vec{k}

Faraday's law also implies that both fields are spatially transverse to each other

$$\vec{B}_0 = \frac{\vec{k} \times \vec{E}_0}{\omega} = \frac{n\hat{n} \times \vec{E}_0}{c} \qquad \hat{n} \equiv \frac{\vec{k}}{k}$$
Here n is the index of refraction.
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Standing Waves

$$\vec{E}(\vec{x},t) = \vec{E}_0 e^{i\vec{k}\cdot\vec{x}-i\omega t}$$
$$\vec{B}(\vec{x},t) = \vec{B}_0 e^{i\vec{k}\cdot\vec{x}-i\omega t}$$

Note that Maxwell's equations are linear, so any linear combination of magnetic/electric fields is also a solution.

Thus a **standing wave** solution is also acceptable, where there are two plane waves moving in opposite directions:

$$\vec{E}(\vec{x},t) = \vec{E}_0 \left(e^{i\vec{k}\cdot\vec{x}-i\omega t} + e^{-i\vec{k}\cdot\vec{x}-i\omega t} \right)$$
$$\vec{B}(\vec{x},t) = \vec{B}_0 \left(e^{i\vec{k}\cdot\vec{x}-i\omega t} + e^{-i\vec{k}\cdot\vec{x}-i\omega t} \right)$$

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Polarization

$$\vec{E}(\vec{x},t) = \vec{E}_0 e^{i\vec{k}\cdot\vec{x}-i\omega t}$$
$$\vec{B}(\vec{x},t) = \vec{B}_0 e^{i\vec{k}\cdot\vec{x}-i\omega t}$$

- As long as the \vec{E} and \vec{B} fields are transverse, they can still have different transverse components.
 - So our description of these fields is also incomplete until we specify the transverse coordinates at all locations in space
 - This is equivalent to an uncertainty in phase of rotation of \vec{E} and \vec{B} around the wave vector \vec{k} .
 - The identification of this transverse field coordinate basis defines the **polarization** of the field.

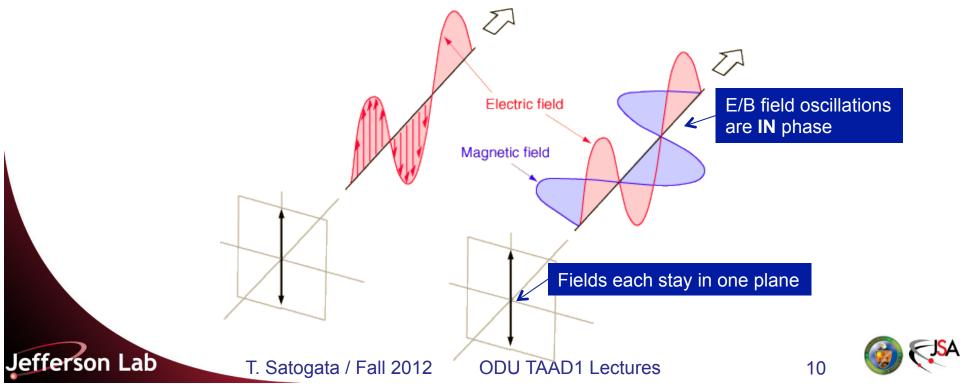
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Linear Polarization

$$\vec{E}(\vec{x},t) = \vec{E}_0 e^{i\vec{k}\cdot\vec{x}-i\omega t}$$
$$\vec{B}(\vec{x},t) = \vec{B}_0 e^{i\vec{k}\cdot\vec{x}-i\omega t}$$

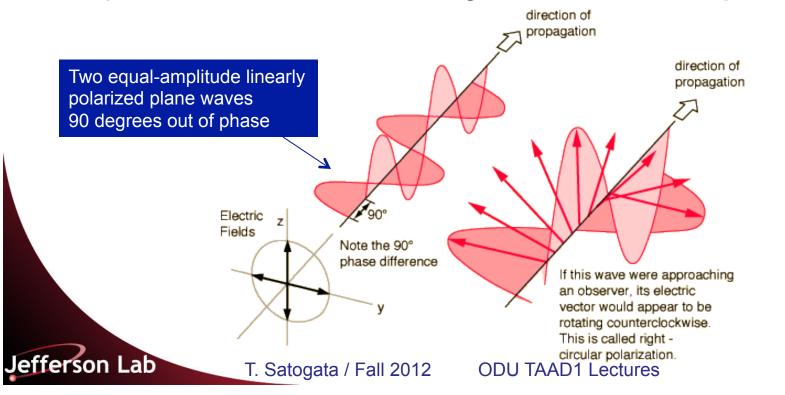
So, for example, if *E* and *B* transverse directions are constant and do not change through the plane wave, the wave is said to be **linearly polarized**.



Circular Polarization

$$\vec{E}(\vec{x},t) = \vec{E}_0 \left(e^{i\vec{k}\cdot\vec{x}-i\omega t} + e^{i\vec{k}\cdot\vec{x}-i\omega t+\pi/2} \right)$$
$$\vec{B}(\vec{x},t) = \vec{B}_0 \left(e^{i\vec{k}\cdot\vec{x}-i\omega t} + e^{i\vec{k}\cdot\vec{x}-i\omega t+\pi/2} \right)$$

• If \vec{E} and \vec{B} transverse directions vary with time, they can appear as two plane waves traveling out of phase. This phase difference is 90 degrees for **circular polarization**.



Back to Maxwell, Now in Accelerators

Not all space is source-free

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- We'll have to deal with boundary conditions
- We'll even have to deal with complicated (linear) combinations of our plane waves in source-free space
- Accelerators have privileged directions
 - E.g. the design momentum direction of the beam
 - This is not always the same as our k direction!
 - So let's break down Maxwell's equations again, now into transverse and longitudinal (z) components



Transverse Separation of Coordinates

- Here \hat{z} labels the direction along the beam, and t labels transverse coordinates perpendicular to \hat{z}

$$\vec{E} = E_{z}\hat{z} + \vec{E}_{t} \qquad \vec{E}_{t} = \vec{E} - (\vec{E} \cdot \hat{z})\hat{z}$$

$$\vec{B} = B_{z}\hat{z} + \vec{B}_{t} \qquad \vec{B}_{t} = \vec{B} - (\vec{B} \cdot \hat{z})\hat{z}$$

$$\begin{bmatrix}\nabla_{t}^{2} + (\mu\varepsilon\omega^{2}) - k^{2}\end{bmatrix}\begin{bmatrix}\vec{E}\\\vec{B} = 0$$

$$\frac{\partial\vec{E}_{t}}{\partial z} + i\omega\hat{z} \times \vec{B}_{t} = \vec{\nabla}_{t}E_{z} \qquad \hat{z} \cdot (\vec{\nabla}_{t} \times \vec{E}_{t}) = i\omega B_{z}$$

$$\frac{\partial\vec{B}_{t}}{\partial z} - i\mu\varepsilon\omega\hat{z} \times \vec{E}_{t} = \vec{\nabla}_{t}B_{z} \qquad \hat{z} \cdot (\vec{\nabla}_{t} \times \vec{B}_{t}) = -i\mu\varepsilon\omega E_{z}$$

$$\vec{\nabla}_{t} \cdot \vec{E}_{t} = -\frac{\partial E_{z}}{\partial z} \qquad \vec{\nabla}_{t} \cdot \vec{B}_{t} = -\frac{\partial B_{z}}{\partial z}$$

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Transverse Separation of Coordinates II

 We can even write down the transverse fields completely in terms of the longitudinal fields

$$\vec{E}_t = \frac{i}{\mu\varepsilon\omega^2 - k^2} \left[k\vec{\nabla}_t E_z - \omega\hat{z} \times \vec{\nabla}_t B_z \right]$$
$$\vec{B}_t = \frac{i}{\mu\varepsilon\omega^2 - k^2} \left[k\vec{\nabla}_t B_z - \mu\varepsilon\omega\hat{z} \times \vec{\nabla}_t E_z \right]$$

Notice how beautifully symmetric these equations are.

This is a helpful decomposition of plane wave solutions to Maxwell's equations into one longitudinal direction and two transverse directions.



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Boundary Conditions: TM and TE Modes

$$\vec{E}_t = \frac{i}{\mu\varepsilon\omega^2 - k^2} \left[k\vec{\nabla}_t E_z - \omega\hat{z} \times \vec{\nabla}_t B_z \right]$$
$$\vec{B}_t = \frac{i}{\mu\varepsilon\omega^2 - k^2} \left[k\vec{\nabla}_t B_z - \mu\varepsilon\omega\hat{z} \times \vec{\nabla}_t E_z \right]$$

 We can come up with some boundary conditions for a couple of simple cases where either electric (TE) or magnetic (TM) fields are completely transverse to
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 Transverse Magnetic (TM)

$$B_z = 0;$$
 Boundary Condition $E_z|_s = 0$

Transverse Electric (TE)

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$$E_z = 0;$$
 Boundary Condition $\frac{\partial B_z}{\partial n}\Big|_S = 0$
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Waveguides

Consider the relationship between the transverse field components

$$\vec{H}_{t} = \frac{\pm 1}{Z} \hat{z} \times \vec{E}_{t}$$
Impedance Z

Wave Impedance

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$$Z = \begin{cases} \frac{k}{\varepsilon \omega} = \frac{k}{k_0} \sqrt{\frac{\mu}{\varepsilon}} & \text{(TM)} \quad B_z = 0\\ \frac{\mu \omega}{k} = \frac{k_0}{k} \sqrt{\frac{\mu}{\varepsilon}} & \text{(TE)} \quad E_z = 0 \end{cases}$$

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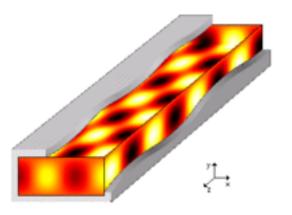
Waveguides

$$\vec{E}(\vec{x},t) = \vec{E}_0 e^{i\vec{k}\cdot\vec{x}-i\omega t}$$
$$\vec{B}(\vec{x},t) = \vec{B}_0 e^{i\vec{k}\cdot\vec{x}-i\omega t}$$

- Plane waves carry electromagnetic energy in the $\ k$ direction and are periodic in all three spatial dimensions
 - We can use that spatial periodicity in transverse boundary conditions, such as at a conductor wall.
 - (Near-)perfect conductors allow us to match transverse boundary conditions

This gives a plane wave propogating longitudinally in a finite transverse volume

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Transverse Fields as an Eigenvalue Problem TM Waves

$$\vec{E}_{t} = \pm \frac{ik}{\gamma^{2}} \vec{\nabla}_{t} \psi \qquad \begin{array}{c} \gamma \equiv (\mu \varepsilon \omega^{2} - k^{2}) \\ \psi \equiv E_{z} \end{array}$$

TE Waves

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$$\vec{H}_{t} = \pm \frac{ik}{\gamma^{2}} \vec{\nabla}_{t} \psi \qquad \gamma \equiv (\mu \varepsilon \omega^{2} - k^{2}) \\ \psi \equiv B_{z}$$

Transverse Helmholz Equation

$$\left(\nabla_{t}^{2}+\gamma^{2}\right)\psi = \left(\frac{\partial^{2}}{\partial x^{2}}+\frac{\partial^{2}}{\partial y^{2}}+\gamma^{2}\right)\psi = 0$$

Differential equation with boundary conditions on E_z or B_z

Boundary Conditions

The boundary conditions are different for TM and TE

$$(\psi \equiv E_z)|_{\rm S} = 0 \ ({\rm TM}) \qquad \left(\frac{\partial(\psi \equiv B_z)}{\partial n}\right)|_{\rm S} = 0 \ ({\rm TE})$$

• For many plane waves of various k values, there are a spectrum of eigenvalues γ_{λ} and eigenfunctions $\psi_{\lambda}(x, y)$ that solve the Helmholz equation

$$\left(\nabla_t^2 + \gamma_\lambda^2\right)\psi_\lambda(x, y) = \left(\frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \gamma_\lambda\right)\psi_\lambda(x, y) = 0$$

Here the wave number \vec{k}_{λ} of a particular mode λ is

$$k_{\lambda}^2 = \mu \varepsilon \omega^2 - \gamma_{\lambda}^2$$

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Cutoff Frequency

 $k_{\lambda}^2 = \mu \varepsilon \omega^2 - \gamma_{\lambda}^2$

- What happens when a mode has a frequency such that the right hand side above is negative?
 - Then k_{λ} is imaginary and the mode does not propagate
 - This property is called cutoff
 - This happens for frequencies below the cutoff frequency

$$\omega_{\lambda}^{2} \equiv \frac{\gamma_{\lambda}^{2}}{\mu\epsilon} \qquad \qquad k_{\lambda}^{2} = \mu\epsilon(\omega^{2} - \omega_{\lambda}^{2})$$

- If you want to propogate a single frequency mode (best for narrow-band response)
 - It's best to build your waveguide so its frequency is below cutoff for the fundamental mode and below cutoff for all other modes